# Synthesis of certified programs in fixed-point arithmetic, and its application to linear algebra basic blocks 

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# Synthesis of certified programs in fixed-point arithmetic, and its application to linear algebra basic blocks 

## Amine Najahi

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Univ. Montpellier 2, LIRMM, UMR 5506
CNRS, LIRMM, UMR 5506


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Floating-point computations
Fixed-point computations

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() Easy and fast to implement
() Easily portable [IEEE754]

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(2) Tedious and time consuming to implement

- $>50 \%$ of design time [Wil98]


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Embedded systems targets

$\mu$-controllers


DSPs


FPGAs
$\rightarrow$ have efficient integer instructions

- Fixed-point arithmetic is well suited for embedded systems

But, how to make it easy, fast, and numerically safe to use by non-expert programmers?

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- Our approach (DALI, Univ. Perpignan):
- certified fixed-point synthesis for:
- Fine grained IP blocks: dot-products, polynomials, ...
- High level IP blocks: matrix multiplication, triangular matrix inversion, Cholesky decomposition



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- High level IP blocks: matrix multiplication, triangular matrix inversion, Cholesky decomposition
- Long term objective: code synthesis for matrix inversion



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How to generate certified fixed-point code for matrix inversion?

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- Contributions:
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- it adheres to the arithmetic model
- Contributions:
- implementation of the arithmetic model



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How to generate certified fixed-point code for matrix inversion?

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2. Build a synthesis tool, CGPE, for fine grained IP blocks:

- it adheres to the arithmetic model
- Contributions:
- implementation of the arithmetic model

3. Build a second synthesis tool, FPLA, for algorithmic IP blocks:

- it generates code using CGPE
- Contributions:

- trade-off implementations for matrix multiplication
- code synthesis for Cholesky decomposition and triangular matrix inversion


## Outline of the talk

1. An arithmetic model for fixed-point code synthesis
2. An implementation of the arithmetic model: the CGPE tool
3. Fixed-point code synthesis for linear algebra basic blocks

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## Fixed-point arithmetic numbers

A fixed-point number $x$ is defined by two integers:
$\triangleright X$ the $k$-bit integer representation of $x$
$\triangleright f$ the implicit scaling factor of $x$

$\rightsquigarrow$ The value of $x$ is given by $x=\frac{X}{2^{f}}=\sum_{\ell=-f}^{k-1-f} X_{\ell+f} \cdot 2^{\ell}$

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How to compute with fixed-point numbers?

## An interval arithmetic based model

- For each coefficient or variable $v$, we keep track of 2 intervals $\operatorname{Val}(v)$ and $\operatorname{Err}(v)$

■ Our model assumes a fixed word-length $k$

## $\operatorname{Val}(v)$ is the range of $v$

## $\operatorname{Err}(v)$ encloses the rounding error of computing $v$

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## $\operatorname{Val}(v)$ is the range of $v$

- the format $\mathbf{Q}_{i, f}$ of $v$ is deduced from $\operatorname{Val}(v)=[\underline{\mathbf{v}}, \overline{\mathbf{v}}]$

$$
\text { - } i=\left\lceil\log _{2}(\max (|\underline{\mathbf{v}}|,|\overline{\mathbf{v}}|))\right\rceil+\alpha \quad \forall f=k-i
$$

$$
\alpha= \begin{cases}1, & \text { if } \bmod \left(\log _{2}(\overline{\mathbf{v}}), 1\right) \neq 0, \\ 2, & \text { otherwise }\end{cases}
$$

## Err(v) encloses the

 rounding error of computing $v$- a bound $\epsilon$ on rounding errors is deduced from

$$
\begin{aligned}
\operatorname{Err}(v) & =[\underline{\mathbf{e}}, \overline{\mathbf{e}}] \\
\cdot \epsilon & =\max (|\underline{\mathbf{e}},|\overline{\mathbf{e}}|)
\end{aligned}
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## An interval arithmetic based model

- For each coefficient or variable $v$, we keep track of 2 intervals Val( $v$ ) and $\operatorname{Err}(v)$

■ Our model assumes a fixed word-length $k$

## $\operatorname{Val}(v)$ is the range of $v$

- the format $\mathbf{Q}_{i, f}$ of $v$ is deduced from

$$
\operatorname{Val}(v)=[\underline{\mathbf{v}}, \overline{\mathbf{v}}]
$$

$-i=\left\lceil\log _{2}(\max (|\underline{\mathbf{v}}|,|\overline{\mathbf{v}}|))\right\rceil+\alpha \quad \forall f=k-i$

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\alpha= \begin{cases}1, & \text { if } \bmod \left(\log _{2}(\overline{\mathbf{v}}), 1\right) \neq 0 \\ 2, & \text { otherwise }\end{cases}
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## Err(v) encloses the

 rounding error of computing $v$- a bound $\epsilon$ on rounding errors is deduced from $\operatorname{Err}(v)=[\underline{\mathbf{e}}, \overline{\mathbf{e}}]$
- $\epsilon=\max (|\underline{\mathbf{e}}|,|\overline{\mathbf{e}}|)$


How to propagate $\operatorname{Val}(v)$ and $\operatorname{Err}(v)$ for $\diamond \in\{+,-, \times, \ll, \gg, \sqrt{ }, /\}$ ?

## Fixed-point multiplication

- The output format of a $\mathbf{Q}_{i_{1}, f_{1}} \times \mathbf{Q}_{i_{2}, f_{2}}$ is $\mathbf{Q}_{i_{1}+i_{2}, f_{1}+f_{2}}$



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- But, doubling the word-length is costly

$-\operatorname{Err}_{\times}=\left[0,2^{-f_{r}}-2^{-\left(f_{1}+f_{2}\right)}\right]$


## Fixed-point multiplication

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- Err $r_{\times}=\left[0,2^{-f_{r}}-2^{-\left(f_{1}+f_{2}\right)}\right]$
- This multiplication is available on integer processors and DSPs

```
int32_t mul (int32_t v1, int32_t v2){
    int64_t prod = ((int64_t) v1) * ((int64_t) v2);
    return (int32_t) (prod >> 32);
}
```


## Our new fixed-point division

- The output integer part of $\mathbf{Q}_{i_{1}, f_{1}} / \mathbf{Q}_{i_{2}, f_{2}}$ may be as large as $i_{1}+f_{2}$



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() sharper bound
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## How to decide of the output format of division?

- A large integer part
$\checkmark$ prevents overflow
$x$ loose error bounds and loss of precision
- A small integer part
$X$ may cause overflow
$\checkmark$ sharp error bounds and more accurate computations


## The propagation rule and implementation of division

■ Once the output format decided $\mathbf{Q}_{i \text { r. } \cdot \text { rr }}$

$-\widehat{\operatorname{Val}\left(v_{2}\right)}=\frac{\operatorname{Val}\left(v_{1}\right)}{\widehat{\operatorname{Val}(v)}+\mathbf{E r r} /} \cap \operatorname{Val}\left(v_{2}\right)$ and $\widehat{\operatorname{Val}(v)}=\left[-2^{i_{r}-1},-2^{-f_{r}}\right] \cup\left[2^{-f_{r}}, 2^{i_{r}-1}-2^{f_{r}}\right]$

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```
int32_t div (int32_t V1, int32_t V2, uint16_t eta)
{
    int64_t t1 = ((int64_t)V1) << eta;
    int64_t V = t1 / V2;
    return (int32_t) V;
}
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```
int32_t div (int32_t V1, int32_t V2, uint16_t eta)
{
    int64_t t1 = ((int64_t)V1) << eta;
    int64_t V = t1 / V2;
    CGPE_ASSERT((()V & 0xFFFFFFFF8000000011) == 0xFFFFFFFF8000000011)
        || ((V & 0xFFFFFFFF8000000011) == 0)));
    return (int32_t) v;
}
```

- Additional code to check for run-time overflows


## The division format trade-off: case of inverting $2 \times 2$ matrices

- Consider $A=\left(\begin{array}{ll}a & b \\ c & d\end{array}\right)$ with $a, b, c, d \in[-1,1]$ in the format $\mathbf{Q}_{2.30}$


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## Outline of the talk

## 1. An arithmetic model for fixed-point code synthesis

2. An implementation of the arithmetic model: the CGPE tool

## 3. Fixed-point code synthesis for linear algebra basic blocks

## The CGPE tool

- CGPE (Code Generation for Polynomial Evaluation): initiated by Revy [MR11]
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2. Filtering step $\rightsquigarrow$ middle-end

- applies the arithmetic model
- prunes the DAGs that do not satisfy different criteria:
- latency $\rightsquigarrow$ scheduling filter
- accuracy $\rightsquigarrow$ numerical filter
- ...



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3. Generation step $\rightsquigarrow$ back-end

- generates C codes and Gappa accuracy certificates



## Code synthesis for an IIR filter using CGPE

- Low-pass Butterworth filter with cutoff frequency $0.3 \cdot \pi$ :

$$
y[k]=\sum_{i=0}^{3} b_{i} \cdot u[k-i]-\sum_{i=1}^{3} a_{i} \cdot y[k-i]
$$

```
<dotproduct inf="0xble91685" sup="0x4e16e97b" integer_width="6" fraction_width="26" width=" 32">
    <coefficient name="b0" value="0x65718e3b" integer_width="-3" fraction_width=" 35" width=" 32"/>
    <variable name="y3" inf="0xble91685" sup="0x4el6e97b" integer_width="6" fraction_width="26" width="32"/>
</dotproduct>
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</dotproduct>
```



## Code synthesis for an IIR filter using CGPE

- Low-pass Butterworth filter with cutoff frequency $0.3 \cdot \pi$ :

$$
y[k]=\sum_{i=0}^{3} b_{i} \cdot u[k-i]-\sum_{i=1}^{3} a_{i} \cdot y[k-i]
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```
<dotproduct inf="0xble91685" sup="0x4e16e97b" integer_width="6" fraction_width="26" width=" 32">
    <coefficient name="b0" value="0x65718e3b" integer_width="-3" fraction_width=" 35" width=" 32"/>
    <variable name="y3" inf="0xble91685" sup="0x4e16e97b" integer_width="6" fraction_width="26" width="32"/>
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```




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## Outline of the talk

## 1. An arithmetic model for fixed-point code synthesis

2. An implementation of the arithmetic model: the CGPE tool
3. Fixed-point code synthesis for linear algebra basic blocks

## A strategy to synthesize code for matrix inversion

## Let $M$ be a matrix of fixed-point variables,

to generate certified code that inverts $M^{\prime} \in M$ a symmetric positive definite, we need to:

1. Generate certified code to compute $B$ a lower triangular s.t. $M^{\prime}=B \cdot B^{T}$
2. Generate certified code to compute $N=B^{-1}$
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- Certified code synthesis for Cholesky decomposition
- Certified code synthesis for triangular matrix inversion
- Certified code synthesis for matrix multiplication



## Linear algebra basic blocks



## Linear algebra basic blocks



## Cholesky decomposition and triangular matrix inversion

Cholesky decomposition

$$
\begin{gathered}
\qquad b_{i, j}= \begin{cases}\sqrt{c_{i, i}} & \text { if } i=j \\
\frac{c_{i, j}}{b_{j, j}} & \text { if } i \neq j\end{cases} \\
\text { with } c_{i, j}=m_{i, j}-\sum_{k=0}^{j-1} b_{i, k} \cdot b_{j, k}
\end{gathered}
$$

Triangular matrix inversion

$$
n_{i, j}= \begin{cases}\frac{1}{b_{i, i}} & \text { if } i=j \\ \frac{-c_{i, j}}{b_{i, i}} & \text { if } i \neq j\end{cases}
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Dependencies of the coefficient $b_{4,2}$ in the decomposition and inversion of a $6 \times 6$ matrix.

## FPLA (Fixed-Point Linear Algebra)



## Impact of the output format of division

## Different functions to set the output format of division

1. $f_{1}\left(i_{1}, i_{2}\right)=t$,
2. $f_{2}\left(i_{1}, i_{2}\right)=\min \left(i_{1}, i_{2}\right)+t$,
3. $f_{3}\left(i_{1}, i_{2}\right)=\max \left(i_{1}, i_{2}\right)+t$,
4. $f_{4}\left(i_{1}, i_{2}\right)=\left\lfloor\left(i_{1}+i_{2}\right) / 2\right\rfloor+t$,
$i_{1}$ and $i_{2}$ : integer parts of the numerator and denominator and $t \in[-2,8]$


Maximum errors with various functions used to determine the output formats of division.

## How fast is generating triangular matrix inversion codes?

- We use $f_{4}\left(i_{1}, i_{2}\right)=\left\lfloor\left(i_{1}+i_{2}\right) / 2\right\rfloor+1$ to set the output format of division


Generation time for the inversion of triangular matrices of size 4 to 40 .

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$\square$ We use $f_{4}\left(i_{1}, i_{2}\right)=\left\lfloor\left(i_{1}+i_{2}\right) / 2\right\rfloor+1$ to set the output format of division


Error bounds and experimental errors for the inversion of triangular matrices of size 4 to 40 .

## Decomposing some well known matrices

- 2 ill-conditioned matrices: Hilbert and Cauchy
- 2 well-conditioned matrices: KMS and Lehmer



## Decomposing some well known matrices

- 2 ill-conditioned matrices: Hilbert and Cauchy
- 2 well-conditioned matrices: KMS and Lehmer


- III-conditioned matrices tend to overflow more often
- similar behaviour in floating-point arithmetic
- The decompositions of KMS and Lehmer are highly accurate


## Conclusions and perspectives

## Contributions

- Formalization and implementation of an arithmetic model
- allows certification
- handles $\sqrt{ }$ and /


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## Perspectives

- Integrate the matrix inversion flow



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