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# A Bulk Built-in Sensor for Detection of Fault Attacks

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**Abstract**—This work presents a novel scheme of built-in current sensor (BICS) for detecting transient fault-based attacks of short and long duration as well as from different simultaneous sources. The new sensor is a single mechanism connected to PMOS and NMOS bulks of the monitored logic. The proposed protection strategy is also useful for improving any state-of-the-art Bulk-BICS from pairs of PMOS and NMOS sensors to single sensors.

**Keywords:** Built-in current sensors, concurrent detection, fault attacks, fault tolerance, security, soft errors, transient faults

## I. INTRODUCTION

Integrated systems based on ultra-deep submicron technologies require higher robustness to natural aging processes or radiation environmental sources [1]. In addition to these natural phenomena, malicious fault-based attacks can be used to bypass security mechanisms of secure systems and to extract information from confidential data [2]. Both these natural or malicious phenomena on integrated circuits can induce transient effects that provoke bit-flips of stored results during the system lifetime. Until the early 2000's, researches about how to mitigate these issues were focused mostly on concurrent error detection and/or correction mechanisms to cope with soft errors induced by transient faults in memory cells. In the last decade, however, more sensitive nanotechnologies as well as the increasing demand for secure systems have also pushed for the development of countermeasures against transient faults in combinational parts of the circuits. These faults indeed can propagate up to storage elements and thus cause soft errors as well. On the other hand, if the transient fault does not induce any error due to an electrical, logical or latching-window masking effect, its detection is crucial all the same in secure applications since the fault itself reveals an attempt of attack.

Among the several strategies to detect transient faults caused by radiation or optical sources, Bulk Built-In Current Sensors (BBICS) offer a promising solution that is perfectly suitable for system design

flows based on CMOS standard cells of commercial libraries. BBICS have the high detection efficiency of costly fault-tolerance schemes (e.g. duplication with comparison) with the low area and power overheads of less efficient mitigation techniques such as time redundancy approaches [3][4][5][6]. Without impacting on system operating frequency, BBICS deal with transient faults of short and long duration as well as multiple faults. Furthermore, as they closely monitor the zones where the faults arise, an early detection is possible immediately after a fault occurrence, preventing induction and propagation of errors to other clock cycles or system blocks.

All today's BBICS-based strategies [5][6][7][8][9][10] use two different circuits that monitor independently PMOS and NMOS bulk currents from pull-up and pull-down CMOS networks. We present in section III of this paper a new lower-area BBICS scheme compounded of a single circuit that is able to monitor at the same time both CMOS networks. Important preliminary related information regarding transient faults and built-in current sensors is initially discussed in section II.

## II. FUNDAMENTALS

### A. Transient Faults in Integrated Circuits

Transient voltage variations during the lifetime of combinational or sequential circuits are defined as transient faults. The first harmful effects of transient faults are soft errors by inverting stored results of system operations (i.e. bit-flips of storage elements). As consequences of transistor shrinking and growing communication of confidential data, soft errors can happen today even at ground level by means of perturbation events arisen from environmental or intentional sources. Examples of environment events are alpha particles released by radioactive impurities and more importantly neutrons from cosmic rays [1].

On the other hand, intentional perturbation events are usually produced by optical sources such as flashlights or laser beams [2] that can maliciously induce transient effects on secure circuits like smartcards to retrieve their secret information.

Soft errors and transient faults are also known as single event upsets (SEU) and single event transients (SET) in fault-tolerance-related fields. If provoked by malicious fault-based attacks, such circuit misbehaviors provide fundamental information for cryptanalysis methods that are able to break security applications.

### B. BICS Detecting Transient Faults

Built-In Current Sensors (BICS) were initially proposed as a mechanism for detecting large increases in the current  $I_{DDQ}$  consumed by a CMOS circuit during its quiescent state, i.e. when the circuit is not switching. The mechanism allows thus testing CMOS circuits against permanent faults [11]. Further, BICS were also adapted for detecting transient faults in memory cells (i.e. bit-flips) [12][13][14][15]. Recently, efforts were made for monitoring transient faults in combinational logic as well [16]. All these techniques connect BICS to the power lines ( $V_{DD}$  and GND) of the monitored circuit to distinguish anomalous transient currents from normal currents. Today's problem is the amplitude of transient currents induced by radiation effects or fault attacks have the same order of magnitude than currents normally generated by switching activities in combinational logic circuits. Hence, schemes monitoring power lines are very limited for detecting just small range of transient faults.

On the other hand, a wide range of transient faults is detectable whether BICS are connected to the bulks of the monitored circuit's transistors; such a smart idea was introduced for the first time in [7][17]. Fig. 1 (a) and (b) illustrate Bulk BICS (BBICS) identifying anomalous transient currents  $I_{fault}$  that flow through the junction between a bulk and a reversely biased drain of a disturbed transistor (MOSFETs "off" in Fig. 1). BBICS indeed take advantage of two facts: (1) In fault-free scenarios (i.e.  $I_{fault} = 0$ ) the bulk-to-drain (or drain-to-bulk) current  $I_B$  is negligible even if the MOSFET is switching in function of new input stimuli; (2) During transient-fault scenarios,  $I_{fault}$  is much higher than the leakage current flowing through the junction. The range of detectable transient faults is

easily adjustable by calibrating the size of the transistors that constitute the BBICS. Hence, schemes based on BBICS can be designed to latch a flag of fault indication for abnormal currents within a defined range that represents a risk of resulting in soft errors.

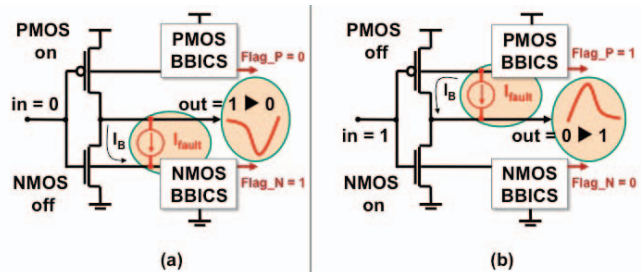


Figure 1. The two classic cases of transient faults in a CMOS inverter perturbed by an anomalous current  $I_{fault}$

Note in Fig. 1 that the connections between monitored circuit (e.g. inverter) and PMOS and NMOS BBICS blocks are done by using metal – from the body-ties of each monitored transistor (i.e. transistor bulks of the inverter) up to the inputs of the BBICS circuitries. Thereby, peaks of anomalous transient currents (i.e. transient faults) are almost not attenuated [18], ensuring thus enough signals for BBICS performing an efficient detection. The only small attenuation is a function of the local distance between the struck zone of the monitored transistor and its body-tie.

## III. NEW SINGLE BBICS

### A. Strategy for Protecting a System

The classic BBICS-based strategy is using a pair of sensors (PMOS and NMOS) to monitor pull-up and pull-down networks – e.g. PMOS-BBICS 1 and NMOS-BBICS 1 in Fig. 2 (a). The capacity of each sensor for identifying transient faults is limited to a certain number of monitored transistors. Hence the strategy is also splitting the target system into blocks with the maximum possible number of transistors the sensor is able to monitor. Fig. 2 (a) illustrates an example of system (chains of inverters) divided into two blocks that are monitored by two pairs of BBICS. For the same example, our strategy is otherwise to use single sensors for pull-up and pull-down networks as Fig. 2 (b) shows.

### B. Structure of Sensor

The circuit structure of any BBICS is composed of two main parts. One is a memory element (e.g. a

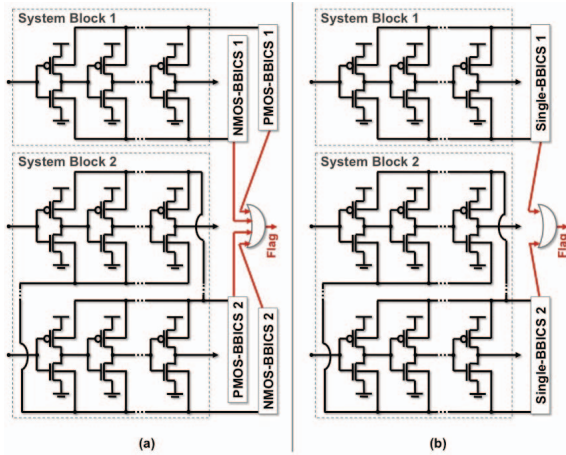


Figure 2. (a) The classic BBICS-based strategy with pairs of BBICS (NMOS and PMOS) monitoring blocks of a system; and (b) Our strategy with the new single BBICS able to check both pull-up and pull-down networks

latch) responsible for registering a flag in case of fault. The other part is a set of transistors that constitute the input circuit of the sensor. It receives the metal connections coming from the body-ties of each monitored transistor. In function of the type of input circuit we classify existing BBICS structures into two groups: (1) *High fault-detecting capacity sensors* [6][7][8] just slightly modify the bulk capacitance of the monitored transistors by using a small input circuit connected to them; thereby these sensors are more capable to recognize a wider range of transient faults. On the contrary, they create an offset on the bulk voltage of the monitored transistors at normal operation (i.e. in fault-free scenarios). It results in slightly lower threshold voltage that represents higher sub-threshold leakage currents [6]; and (2) *Low sub-threshold leakage sensors* [5][7][9][10] avoid any voltage offset on the bulk by connecting it via a high ohmic transistor (large channel length) in on-state with  $V_{dd}$  or  $gnd$ . On the other hand, they considerably increase the bulk capacitance and, thus, decrease their fault-detecting capacities.

The structure of the new single BBICS presented in this paper takes the high fault-detecting capacity of the sensors proposed in [6] and illustrated in Fig. 3 (a). Moreover, it was optimized from a pair of sensors to a single sensor and, consequently, the area could be significantly reduced as we can see in Fig. 3 (b) compared with Fig. 3 (a) in terms of transistor counting. Low leakage power overheads can be met by finding an optimal number of monitored transistors. In Fig. 3,  $W_{min}$  represents the minimum

diffusion width of the transistors,  $L_{min}$  is the minimum channel length, and design factors  $Xn$ ,  $Yn$ ,  $Xp$ ,  $Yp$ ,  $X$  and  $Y$  are used for calibrating the sensor with a desired fault-detecting capacity.

Transistors 5, 6, 7, and 8 in Fig. 3 (b) constitute the latch used to memorize a flag (i.e. output *Flag* at  $V_{dd}$ ) in case of a transient fault within a defined range of current that temporally perturbs the monitored pull-up or pull-down network. In addition, the latch transistors are responsible for amplifying the anomalous transient currents coming from node *NMOS\_Bulk* or going to node *PMOS\_Bulk* of monitored blocks. Factors  $X$  and  $Y$  allow increasing the gain of amplification and, thus, configuring the sensor with higher fault-detecting capacity. Transistors 11 and 12 are used to set the sensor in sleep-mode when the system is left on standby (i.e. signal *Sleep\_Mode* at  $V_{dd}$ ). It leads to zero the offset on the bulk voltage of the monitored block, and so the sub-threshold leakage power consumption is significantly reduced.

### C. Mode of Operation

In fault-free conditions and inactive sleep mode (i.e. *Sleep\_Mode* at  $gnd$ ), the proposed structure in Fig. 3 (b) has in theory *NMOS\_Bulk* at  $gnd$  level and *PMOS\_Bulk* at  $V_{dd}$ . Signal *Flag* is at  $gnd$ ; *node1* at  $V_{dd}$ ; and transistors 1, 3, 4, 10, 11, and 12 are in off-state. Further, as transistors 2 and 9 are in on-state, they impose, in reality, on *NMOS\_Bulk* and *PMOS\_Bulk* a small voltage drop.

In case of a transient fault in the pull-up network of the monitored block, *PMOS\_Bulk* acts as a current sink, i.e. a transient current flows from *node1* to *PMOS\_Bulk* through transistor 9. Thereby the voltage of *node1* decreases, leading the inverter formed by transistor 6 and 5 to flip and signal *Flag* to change from  $gnd$  to  $V_{dd}$ . As consequence, the other inverter (i.e. transistor 7 and 8) also switches forcing *node1* to  $gnd$  even after the transient fault has ceased. Moreover, it activates transistors 10 and 1 in order to set the normal operation levels for *PMOS\_Bulk* and *NMOS\_Bulk* (i.e.  $V_{dd}$  and  $gnd$ ) during such a state of fault indication at *Flag*. As soon as this flag is processed by higher instances of the system, it must be reset (i.e. input *Reset* at  $V_{dd}$ ) to be able to detect again other transient faults.

If a fault happens in the pull-down network of the monitored block, a transient current flows through

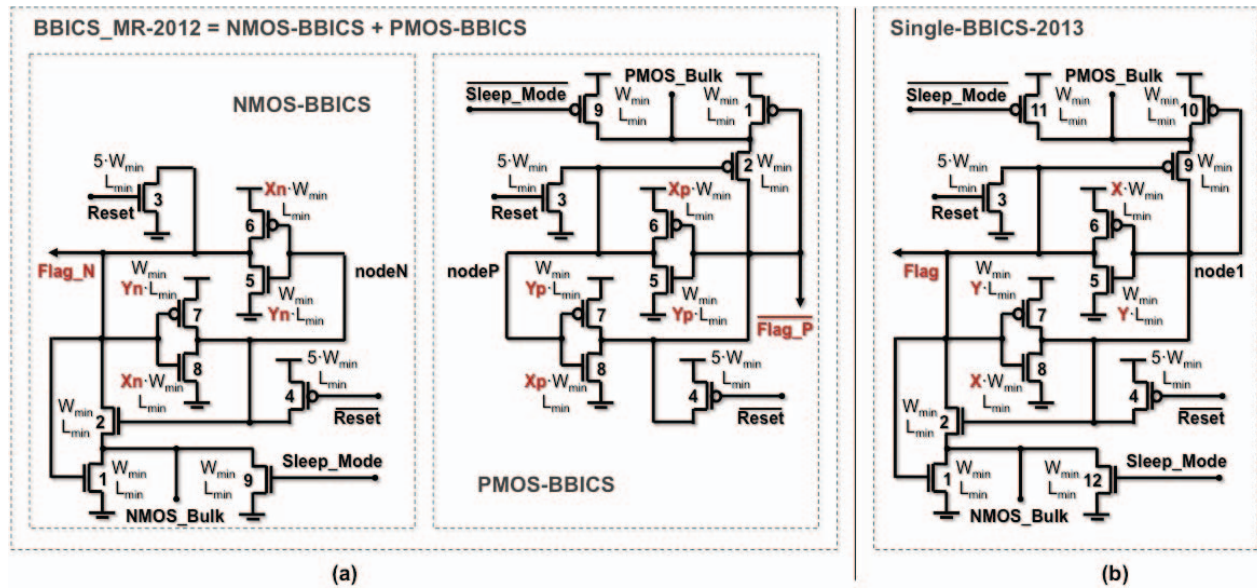


Figure 3. (a) Structures of a BBICS pair (NMOS and PMOS) proposed in [6]; and (b) the structure of our new single BBICS

transistor 2 coming from *NMOS\_Bulk* and, consequently, the same procedure described above is performed to register an indication of fault (i.e. signal *Flag* from *gnd* to *V<sub>dd</sub>*). Note that if other transient fault occurs simultaneously in pull-up network, the single BBICS is able to accomplish the memorization of a flag since the sense of the positive feedback is the same for both fault events. On the contrary, no new transient fault can be detected when a flag of fault is latched or during the reset of the sensor.

#### IV. CONCLUSIONS

With a count of only 12 transistors for monitoring PMOS and NMOS networks, the new BBICS proposed in this paper presents lower area than previous versions. No additional logic is required to keep, refresh, or gather the fault indications from CMOS networks. Moreover, the number of total transistors can be reduced to 9 by eliminating one of the reset transistors and the sleep mode. Ongoing works are the fabrication of a prototype and the validation of the approach using a laser beam.

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