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A New Augmented $\mathcal{L}_1$ Adaptive Control for Wheel-Legged Robots: Design and Experiments

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Abstract—This paper proposes the augmentation of an $\mathcal{L}_1$ adaptive controller with a feedback Linear Quadratic Regulator (LQR) to control a wheel-legged biped robot. The performance of linearized model-based controllers, such as LQR, depends on the accurate knowledge of model parameters, a priori information about input and output disturbances, and other unforeseen conditions. We propose a hybrid scheme where an $\mathcal{L}_1$ adaptive controller is combined with LQR to compensate for matched uncertainties and other disturbances related to the environment change such as friction conditions of the floor. The proposed control scheme is able to keep the robot stable under model uncertainties and external disturbances through a series of validation scenarios including simulations and real-time experiments.

I. INTRODUCTION

The high maneuverability, speed, and agility of self-balancing wheel-legged robots compared to the legged humanoid robots come in exchange for instability and challenging control system design. Due to a small footprint and a tall body, these robots are well suited and even desired for indoor settings such as restaurants, banks, homes, hotels, etc. However, the robust self-balancing of wheel-legged robots remains the most challenging aspect for deploying these robots in the real world. Most of these applications require a robot that may carry an object from one place to other for automated delivery, or the human guidance service by moving on different floor conditions. These practical requirements further enhance the difficulty in the control problem as the unknown payload incorporates model uncertainty, and the unknown floor friction also creates the need for environmental adaptation.

The safety-critical nature of these systems requires a robust control system design for all practical purposes, and hence gained a significant interest among the control and robotics communities in recent years.

A. Background

In the literature, wheeled inverted pendulum (WIP) robots are divided into two categories, namely with legs also known as wheel-legged robots and without legs. Earlier works on WIP robots were mostly focused on the version without legs. Yuta et al. proposed pitch and trajectory control of a WIP robot called Yamabico Kurara and successfully separated its steering control from balancing and translational motion [1] [2]. Tani et al. attempted cooperative object transportation using an unstable WIP robot and a human [3]. A control system was built to estimate the external force and exert the required force to maintain the robot in a balanced state.

A team from EPFL developed a WIP robot called JOE [4]. They implemented two decoupled schemes to control its orientation and translational motion along with balancing. In another important study, Agrawal and his team produced partial feedback linearization of a WIP robot while considering its nonholonomic constraints [5] [6]. Takahashi et al. introduced an assistant robot with wheeled inverted pendulum mechanism capable of various tasks such as standing, sitting, as well as object picking [7] [8]. A linear quadratic regulator (LQR) was devised and applied for its motion control.

More recently, Boston Dynamics announced its first wheel-legged robot named Handle that can be used as a pick and place robot in industrial warehouses [9]. Also a team from Harbin Institute of technology in China designed a hoseless hydraulic wheel-legged robot able to improve the overall reliability of the hydraulic actuator system [10].

A team from ETH, Zurich also developed a wheel-legged robot called Ascento [11] [12]. The control scheme includes a whole-body controller that incorporates rolling constraints for better performance against curves and an LQR used for the pitch control. Caporale et al. proposed a computed torque control law to stabilize a wheeled humanoid robot [13].

(a) Real robot. (b) 3D model in ROS Rviz.

Fig. 1: View of the Igor wheel-legged robot.
Machine learning techniques are also getting attention recently for balancing and control of wheel-legged robots. In a very recent study, Jiang et al. proposed a novel data-driven value iteration algorithm that generates a balancing controller for a wheel-legged robot with a small amount of data [14]. Model learning of a two-wheeled robot where a simulation model is learned first and then the difference model to reduce the sim-to-real gap has also been reported [15]. Further studies also proposed nonlinear sliding-mode controller, optimization based whole-body controller, and nonlinear optimal control for WIP robots [16] [17] [18] [19].

B. Motivation

It is clear from the previous discussion that both linear and nonlinear motion controllers have been designed to balance and control WIP robots. Commonly used model-based controllers depend on system parameters which may change over time, and other assumptions to simplify the system for modeling purposes. Moreover, linear motion controllers such as LQR are mainly based on linearized models of the system and hence relevant for only a small region around the operating point. We believe an adaptive control mechanism is essential for the wheel-legged system under the assumed circumstances. Accordingly, we propose a new control scheme based on the combination of an LQR feedback controller and an $\mathcal{L}_1$ adaptive controller to compensate for modeling errors, external disturbances, and other eventual uncertainties. To the best of the authors’ knowledge, it is the first time that an adaptive control scheme has been proposed for a wheel-legged robot with experimental validation on a real system.

II. MODEL OF THE WHEEL-LEGGED ROBOT

This section provides the mathematical modeling of the wheel-legged robot named Igor illustrated in Fig. 1. The robot has three degrees of freedom (DoF); including the translational motion, the heading angle or orientation, and the pitch angle. Three reference frames $\Sigma_I$, $\Sigma_R$, and $\Sigma_C$ are considered to completely describe the robot in the world frame as illustrated in Fig. 2. $\Sigma_I$ is the inertial frame attached to the ground, $\Sigma_R$ is the robot base frame, and $\Sigma_C$ is the Center of Mass (CoM) frame. Further, the origins of $\Sigma_R$ and $\Sigma_C$ coincide.

The pitch angle ($\beta$) is defined as the angle between $Z_R$ and $Z_C$, representing Z-axes of $\Sigma_R$ and $\Sigma_C$, respectively. The heading angle ($\alpha$) represents the angle between $X_R$ and $X_I$ axes of $\Sigma_R$ and $\Sigma_I$ frames. Six parameters are used to derive the dynamics of the robot in the inertial frame; $\beta R$, $\alpha R$, and $\theta I$ define the robot base location, $\theta_R$ and $\theta_I$ denote the right wheel and left wheel displacements, respectively, and $\alpha$, $\beta$ are the yaw and pitch displacements of the robot.

The nonlinear dynamic model of the robot is then obtained as

$$\dot{q} = \frac{\mathbf{J}(q)}{M(q)} \ddot{\mathbf{q}} + \mathbf{Y}(q, \dot{q}) + \mathbf{G} + \mathbf{E} \dot{\lambda} + \mathbf{J}(q)^T \lambda,$$

(1)

where $M(q) \in \mathbb{R}^{6 \times 6}$ is the inertia matrix, $\mathbf{V} \in \mathbb{R}^{6 \times 6}$ is a matrix of viscous coefficients, $\mathbf{H}(\dot{q}, \dot{\mathbf{q}}) \in \mathbb{R}^6$ includes centrifugal and coriolis terms, $\mathbf{G} \in \mathbb{R}^6$ is the gravity vector, $\mathbf{E} \in \mathbb{R}^{6 \times 2}$ is the torque selection matrix, $\mathbf{T} \in \mathbb{R}^2$ is the control input torque vector, $\mathbf{J}(q)$ defines nonholonomic and holonomic constraints of the robot, $\lambda$ is Lagrange multiplier, and vector $q = [\beta R \alpha R \theta I \theta R \alpha \beta \theta I]^T$ represents the generalized coordinates.

Finally, after removing the Lagrange multiplier $\lambda$ and linearizing the model around its equilibrium point $\beta = 0$, we get the reduced-order linear model in the state-space form as follows

$$\dot{X} = AX + BU$$

$$Y = CX + DU,$$

(2)

where $X = [p \alpha \beta \dot{\alpha} \dot{\beta}]^T$ is the state vector, and $p = \beta R \cos(\alpha) + \dot{\alpha} \sin(\alpha)$ is the translational position of the robot, $Y \in \mathbb{R}^6$ represents the output vector of the system, and $U \in \mathbb{R}^2$ is the control input. The matrices $A$, $B$, $C$, and $D$ are respectively called state, input, output, and feedforward matrices.

III. PROPOSED CONTROL SCHEME

This section introduces the proposed control method to stabilize the wheel-legged robot which is inherently unstable in nature. The proposed augmented $\mathcal{L}_1$ adaptive control
A linear quadratic regulator (LQR) is a full state feedback optimal controller that places the poles of the closed-loop system to minimize the following quadratic cost function

\[ J_{lqr} = \int_0^\infty [X(t)^T Q X(t) + U(t)^T R U(t)] \, dt, \]

where \( Q \) and \( R \) define weights for the system states and the control inputs, respectively.

The state feedback control law that minimizes the above cost function is

\[ U = r e f - K_{lqr} X, \]

where \( r e f \) represents the reference signal. The state feedback gain \( K_{lqr} \) is given by

\[ K_{lqr} = R^{-1} B^T P, \]

where \( P \) is obtained by solving the following algebraic Riccati equation (ARE)

\[ A^T P + PA - PBR^{-1}B^T P + Q = 0 \]

The matrices \( A \) and \( B \) are obtained using the parameters of the wheel-legged robot summarized in TABLE I, while \( Q \) and \( R \) are tuned with a trial-and-error method to get satisfying performance in the real-time system. After obtaining the matrices, we obtained the following \( K_{lqr} \) value by utilizing the pole placement technique,

\[ K_{lqr} = \begin{bmatrix} -2.8284 & 1.4142 & -19.4979 & -4.8849 & 0.0432 & -4.7839 \\ -2.8284 & -1.4142 & -19.4979 & -4.8849 & -0.0432 & -4.7839 \end{bmatrix}. \]

B. Background on \( \mathcal{L}_1 \) Adaptive Control

This section describes the design of an adaptive control system for the pitch angle \( (\beta) \) stabilization of the wheel-legged robot. For a safety-critical system like a self-balancing robot, keeping it in the upright position is of utmost importance to avoid any accident, and damage of the robot and its environment. Accordingly, we introduce the method where an \( \mathcal{L}_1 \) adaptive controller is augmented with an LQR controller in order to keep the robot stable in unforeseen circumstances.

The decoupling between adaptation and robustness ensured by the \( \mathcal{L}_1 \) adaptive control architecture makes it an ideal adaptive controller for real-time applications. High adaptation gains for achieving fast convergence can be used without introducing a high frequency signal in the control input. The control scheme proposed in this section is based on the \( \mathcal{L}_1 \) adaptive control theory for systems with time-varying parameters and disturbances along with uncertain system input gain [22].

The \( \mathcal{L}_1 \) adaptive control consists of an adaptation and a prediction stage as illustrated in Fig. 4. The adaptation phase is used to predict the unknown and/or time-varying parameters and other uncertainties including external disturbances, whereas the prediction stage is used to get the ideal required performance of the system. Furthermore, a low pass filter is incorporated in the closed-loop to remove high frequencies from the control signal that may occur due to high adaptation gains.

Let us consider the inverted pendulum model that is extracted from (1) in the form

\[ \dot{x}(t) = A_p x(t) + B_p (\hat{\omega}_{ad}(t) + \hat{\sigma}^T(t) x(t) + \sigma(t)), \]

\[ y(t) = C_p x(t) \]

where \( A_p \) is the known \( \mathbb{R}^{2 \times 2} \) matrix that describes the linear dynamics of the inverted pendulum, \( x(t) = [\beta \dot{\beta}]^T \in \mathbb{R}^2 \) is the state vector, \( B_p \in \mathbb{R}^2 \) and \( C_p \in \mathbb{R}^{2 \times 2} \) known matrices, \( \theta(t) \in \mathbb{R}^2 \) vector of time-varying unknown parameters, \( \sigma(t) \in \mathbb{R} \) models eventual disturbances and unmodeled dynamics, \( \Omega \in \mathbb{R} \) is an unknown positive constant and \( u_{ad}(t) \in \mathbb{R} \) is the adaptive control input.

1) State Predictor: To develop a full-state feedback controller so that the system output \( y(t) \) tracks the reference signal \( r(t) \), we consider a state predictor of the form

\[ \dot{x}(t) = A_m \dot{x}(t) + B_p (\hat{\Omega}(t) u_{ad}(t) + \hat{\theta}^T x(t) + \hat{\sigma}(t)), \]

\[ \dot{y}(t) = C_p \dot{x}(t) \]

where \( A_m \) is \( A_p - B_p K_m \in \mathbb{R}^{2 \times 2} \) is a known Hurwitz matrix, whereas \( \hat{\theta}, \hat{\sigma} \), and \( \hat{\Omega} \) are the estimates of \( \theta, \sigma \), and \( \Omega \), respectively.

2) Adaptation Law: The estimations of the parameters are governed by the following projection-based adaptive laws

\[ \dot{\hat{\theta}}(t) = \text{Proj}(\dot{\hat{\theta}}(t), -\Gamma_\theta \ddot{x}(t) P_a B_p x(t)), \]
\[ \dot{\hat{\sigma}}(t) = \text{Proj}(\dot{\hat{\sigma}}(t), -\Gamma_\sigma \ddot{x}(t) P_a B_p), \]
\[ \dot{\hat{\Omega}}(t) = \text{Proj}(\dot{\hat{\Omega}}(t), -\Gamma_\Omega \ddot{x}(t) P_a B_p u_{ad}(t)), \]

\[ \dot{\hat{\Theta}}(0) = 0, \]

\[ \dot{\hat{\Theta}}(t) = \dot{\hat{\Theta}}(t) - x(t) \]

(7)

where \( \Gamma_\Theta > 0, \Gamma_\sigma > 0, \) and \( \Gamma_\theta > 0 \) are the adaptation gains, \( \ddot{x}(t) = \dot{x}(t) - x(t) \) is the prediction error, and \( P_a = P_a^T > 0 \) is the solution of Lyapunov equation \( A_m^T P_a + P_a A_m = -Q_a \) for an arbitrary symmetric matrix \( Q_a = Q_a^T > 0 \).

3) Projection Operator: A projection operator is used for updating the parameters \( \hat{\theta}, \hat{\sigma}, \) and \( \hat{\Omega} \) smoothly and confining them within the required set [23]. The algorithm of the projection operator \( \text{Proj}(z, \phi) \), used in (7) for a parameter \( z \), is described as follows,

![Fig. 4: Block diagram of \( \mathcal{L}_1 \) Adaptive Control.](image)
Algorithm: Projection Operator

Inputs: $\epsilon$, $z$, $\phi$, $z_{\text{max}}$, $z_{\text{min}}$

1: compute $f_d = (z_{\text{max}} - z_{\text{min}})^2$;
2: compute $f_z = \frac{4z_{\text{max}}z_{\text{min}} - 2z_{\text{max}} - 2z_{\text{min}}}{\epsilon f_d}$;
3: compute $f_\phi = \frac{4z_{\text{max}}z_{\text{min}} - 2z_{\text{max}} - 2z_{\text{min}}}{\epsilon f_d}$;
4: define $output = \phi$;
5: if $(f_z <= 0$ and $f_\phi * \phi < 0)$ then $output = \phi * (f_z + 1)$;
6: return output;

here, $0 < \epsilon < 1$ is a constant that sets the steepness of the curve. $z_{\text{max}}$ and $z_{\text{min}}$ are respectively the maximum and minimum values delimiting the admissible range of the parameter $z$.

4) Adaptive Control Law: According to the block diagram of Fig. 4, the adaptive control input $u_{\text{ad}}$ for the inverted pendulum system (5) is given in Laplace domain as follows:

$$u_{\text{ad}}(s) = -K_f D(s) \left( \hat{\eta}(s) - K_g r(s) \right),$$

where $K_g$ is the feedforward gain, $r(s) = [\beta_d, \dot{\beta}_d]^T$ is the reference signal, $K_f > 0$ is the feedback gain, and $D(s)$ is a strictly proper transfer function such that

$$C(s) = \frac{\Omega K_f D(s)}{1 + \Omega K_f D(s)},$$

is strictly proper stable transfer function and DC gain $C(0) = 1$. Furthermore,

$$\dot{\eta}(t) = \hat{\Omega} u_{\text{ad}}(t) + \hat{\theta}^T(t) x(t) + \hat{\sigma}(t).$$

To ensure the stability of the resulting closed-loop system, the design of the feedback gain $K_f$ and the low-pass filter $D(s)$ should satisfy the following $L_1$-norm condition:

$$\|G(s)\|_{L_1} L < 1,$$

where $G(s) = H(s)(1 - C(s))$, $H(s) = (sI - A_m)^{-1} B_g$, and $L = \max_{\theta \in \Theta} \|\theta\|_1$, here $\Theta$ is a known convex compact set.

The reader is encouraged to refer to [22] for detailed proofs of stability and performance analysis.

IV. SIMULATION RESULTS

This section provides an overview of the simulation setup used in this study as well as the corresponding obtained results. The wheel-legged robot Igor was defined in the Unified Robotic Description Format (URDF) and simulated in Gazebo simulator using the Robotic Operatic System (ROS). We used C++ programming language to implement the proposed motion control algorithm for better portability and efficiency. The control loop runs at a frequency of 500Hz. The nominal parameters of the system are summarized in TABLE I. To validate the proposed augmented adaptive control scheme qualitatively as well as quantitatively, we conducted two different tests and results are shown in Fig. 5. The main motivation behind these scenarios is to compare the input torques of both controllers, and the pitch angle ($\dot{\beta}$) and pitch rate ($\ddot{\beta}$) subject to external disturbances and model uncertainties.

In the first scenario, we used the nominal parameters of the robot and compare the augmented control scheme against the model-based LQR. The obtained results from this simulation are depicted in Fig. 5a. We applied a linear force of $20N$ on the robot body at $t = 10s$ for a duration of $1s$. The given plots clearly indicate that the augmented controller keeps the pitch angle ($\beta$) and rate ($\dot{\beta}$) smaller than those of the LQR. Furthermore, the settling time of the pitch angle for the proposed controller is about $2.5s$ against $6s$ for LQR. In the second scenario, to introduce a modeling uncertainty we propose to change the value of the viscous friction coefficient of the wheel actuators from its nominal value $c_r = c_l = 0.17$ to $c_r = c_l = 0.34$ (i.e. $+100\%$). We know that determining these friction coefficients in the real world is near impossible as they change over time due to abrasion and other factors. In case of the model-based LQR as shown in Fig. 5b, it is evident that the impact of inaccurate friction coefficients can be highly risky for a self-balancing robot as it could not recover from the translational push of $20N$. On the other hand, the proposed augmented controller successfully kept the robot around its vertical upright position.

TABLE II summarizes the quantitative differences between the LQR and the proposed augmented controller in different settings. By integrating the torques over time, which is also known as angular impulse, we found that in the nominal case, it is $6.3441$ N.m.s against $3.6355$ N.m.s for the LQR and the proposed augmented controller, respectively. It is clear that the augmented controller uses about $42.7\%$ less torque and hence less energy than the LQR in keeping the robot balanced in the nominal case. Besides it is worth to note that the proposed augmented control scheme can endure a linear push of up to $30N$ i.e. $50\%$ more force than the maximum of $20N$ for the LQR.

V. REAL-TIME EXPERIMENTS AND RESULTS

For the real-time control of the Igor robot, we used ROS with C++ to run the motion control algorithms. The control loop runs on an intel core i7 microprocessor at a frequency of 500Hz and the torque commands are transmitted to the robot actuators through a wifi connection. Besides, for the robot localization we used an extended kalman filter (EKF) to filter and fuse the data from multiple IMUs and the wheel

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Description</th>
<th>Value</th>
<th>Units</th>
</tr>
</thead>
<tbody>
<tr>
<td>$m_c$</td>
<td>Robot body mass.</td>
<td>7.5</td>
<td>Kg</td>
</tr>
<tr>
<td>$m_w$</td>
<td>Robot wheel mass.</td>
<td>0.35</td>
<td>Kg</td>
</tr>
<tr>
<td>$l$</td>
<td>Distance of the CoM from the origin of $\Sigma_R$.</td>
<td>0.5914</td>
<td>m</td>
</tr>
<tr>
<td>$r_w$</td>
<td>Wheel radius.</td>
<td>0.1016</td>
<td>m</td>
</tr>
<tr>
<td>$c_r$, $c_l$</td>
<td>Right wheel and left wheel viscosity coefficients.</td>
<td>0.17 $\frac{N.m}{rad/sec}$</td>
<td></td>
</tr>
</tbody>
</table>
Fig. 5: Validation in simulation: Temporal evolution of pitch angle, pitch rate, wheel torque, and the estimated parameters of the augmented $L_1$ adaptive controller during a push of 20N force.

odometry.

We performed a couple of experiments on the real Igor robot to compare the stability performance and energy efficiency of the two controllers in different operating conditions. The controllers have the regulation task to keep the robot balanced $\beta \approx 0$ in the presence of the external pushes. Since the real robot lacks a force sensor, we applied manual pushes repeated three times to take the average result. The first experiment is performed on a normal tiled floor with less friction between the floor and the wheels of the robot, while the second experiment is performed on a carpet floor with a better friction between the robot wheels and the ground. Also a lidar sensor of mass 1Kg is attached to the bottom of the Igor body for emulating modeling uncertainties. This extra mass is not included in the dynamic model of the robot and the LQR controller. The obtained results of these real-time experiments are displayed in Fig. 6 and further summarized quantitatively in TABLE II.

For comparison, we repeated the same tests on a carpet surface in the second experiment as shown in Fig. 6b. It is noted that with similar pushes, the deviation of the pitch angle from the given reference point stays smaller for the proposed augmented controller than for the LQR. However, this difference is reduced from 59.2% to 36.3% in the case of carpet floor. Furthermore, due to high friction between the carpet floor and the robot wheels, both controllers give the similar torque profiles to keep the robot balanced against the applied external pushes.

VI. CONCLUSION AND FUTURE WORK

In this paper we proposed and implemented a linear quadratic regulator augmented with an $L_1$ adaptive controller to stabilize a wheel-legged robot. It is shown through different simulations and real-time experiments that the proposed augmented controller was successfully able to compensate for the model uncertainties and external disturbances and clearly outperforms the model-based LQR controller.

| TABLE II: We use area under the curve of the pitch angle and the wheel torque to quantify the performance of the given controllers. The smaller these values are the better the corresponding controller performs in terms of settling time, overshoot, and energy consumption. The percent change indicates the percentage reduction of these areas in the case of augmented controller w.r.t the LQR. |
| --- | --- | --- | --- |
| **Simulations** | LQR | Augmented Controller | Percent Improved |
| Nominal case | $\int |\beta|dt$ | 0.6157 | 0.3070 | 50.1% |
| Uncertain case | $\int |\beta|dt$ | 6.3441 | 3.6535 | 42.7% |
| Nominal case | $\int |\tau|dt$ | 11.5773 | 0.5006 | 95.7% |
| Uncertain case | $\int |\tau|dt$ | 18.0505 | 16.2275 | 10.0% |
| **Real-time Experiments** | LQR | Augmented Controller | Percent Improved |
| Tile | $\int |\beta|dt$ | 0.6624 | 0.2700 | 59.2% |
| floor | $\int |\tau|dt$ | 13.5188 | 8.8440 | 34.6% |
| Carpet | $\int |\beta|dt$ | 0.5160 | 0.3283 | 36.3% |
| floor | $\int |\tau|dt$ | 10.9857 | 10.9037 | 0.75% |
In the future work, we may focus on design and implementation of an $L_1$ adaptive controller for the underactuated wheeled-robot to control its further DoFs such as translational position, pitch angle, and the yaw angle.

REFERENCES


